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*Full Length Research Paper*

# Performance Response Characteristics of Avalanche Photodiodes (APDs) Under High Thermal and Protons Irradiation Environments

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This paper has been examined the high temperature irradiation variations testing in order to be used to determine avalanche photodiode lifetime, even though APD failure mechanisms are more sensitive to increases in current density. As a measured parameter of degradation, the current density is of great significance when searching for failure modes in APD. Raising the current density however, is not really indicative of lifetime since it is more likely a situation to be avoided than one that simulates normal lifetime degradation. The reliability of semiconductor detectors is very dependent on the degradation modes. This paper has investigated deeply some of the degradation performance and capabilities of typical APDs currently used in many communication and sensing systems over wide range of the affecting parameters. APDs are used in systems that require coherent and often single mode light such as high data rate communications and sensing applications. APDs are an attractive receiver choice for photon-starved (low signal) applications, because their internal gain mechanism can improve signal to noise ratio. An optical receiver must also be appropriate for the laser wavelength being used. The near infrared is the preferred wavelength regime for deep space optical communications largely due to the wavelengths of available laser technologies that meet the optical power requirements of a deep space optical link.

**Keywords:** Proton radiation damage, Optical functional device, High temperature environments and Radiation effects.

## INTRODUCTION

In several fiber optic communication systems, the avalanche photodiode (APD) is preferred to the p-i-n photodetector (Sze, 2007; Squillante., et al 1986), since it provides high quantum efficiency, high frequency response, low noise, and a higher current gain than a p-i-n diode due to the avalanche multiplication phenomenon. The internal gain mechanism can improve signal to noise ratio, So the coupling of the APD with scintillators

provides the high resolution detection of the high-energy particles and gamma-ray (Moszynski., et al 1999), (Christopher., et al 2007). Also, Silicon avalanche photodiodes are among the most sensitive photodetectors for visible and near infrared wavelengths (Chistokhin., et al 2009; Andrea ., et al 2003; (Arbat., et al 2010) . Hence, it is important to study the effects of nuclear radiation on the performance of these devices.

The study of nuclear radiation effects on semiconductors shows that two types of defects are introduced. They are ionization damage and displacement damage. Both ionization defects and displacement damage could lead to permanent damage of the semiconductor material (Geoffrey ., et al 1993). But ionization damage are mostly transient and usually causes little permanent damage to the photodiode performance for total doses below  $10^5$  rad (Si) [(Shinobu ., et al 2006: (Cappelletti1., et al 2006 : Cappelletti., et al 2008 : Kwang ., et al 2008). Previous studies of radiation damages of APDs used either gamma rays, electrons, protons or neutrons as the radiation sources (Swanson., et al 1987: Dai., et al 1996: Grahl., et al 2003: Ikagawa., et al 2005: Pulvirenti., et al 2008: Ogasawara., et al 2010) . Gamma rays were believed to cause the same amount of ionization damage as protons of the same dose but very little displacement damage (Antunovic., et al 2005). Neutrons cause only displacement damages. One may estimate the displacement damage due to protons by scaling the neutron radiation damage (Xiaoli Sun and Henri Dautet, 2001: Antunovic., et al 2005: Ruzin., et al 1999). Ionization damage is the major concern for surface devices like MOSFET's. APD's are junction devices and are much less sensitive to ionization damages. Although ionization damage can cause significant increase in the APD surface leakage current due to the increase of surface states, its contribution to the total output noise is often negligible compared to the bulk leakage current. Displacement damage (DD) effectively adds defect sites to the semiconductor crystal, resulting in a rise in the bulk dark current and excess noise factor of APDs fabricated in Si [(Pilicer., et al 2005), (Gedam., et al 1991) and InGaAs-InP (Laird., et al 2005), (Jamie ., et al 2003). Besides an increase in dark current, which obviously reduces photocurrent as defects act as electron or hole trapping centers for the photogenerated pairs, also increases the Noise Equivalent Power by increasing shot noise, DD can also alter the effective doping level structure by either donor removal and/or acceptor compensation (Lindstrom., et al 2006), both of which can alter the quantum efficiency and gain. For example, previous reports by Osborne et al. (Baccaro., et al 1999) and Baccaro et al. (Osborne., et al 2000) found neutron irradiation to levels of around  $10^{13}$  n/cm resulted in gain degradation (Ercan and Tapan, 2007). In some instances, radiation damage also leads to a reduction in the breakdown bias as observed in some Si [(Ikagawa., et al 2005), (Jamie ., et al 2006), (Jamie ., et al 2006) and InGaAs devices [(Laird., et al 2005), (Jamie ., et al 2003), (Becker, 2004). Indeed some authors have indicated that damage can even result in catastrophic failure (Pilicer., et al 2005), (Ingram., 2010) via a mechanism which depends on the reliability of the near surface region.

The evolution of optoelectronic materials and fabrication methods is the most important factor in interpreting older data because all of the older work was done on part technologies that are so different from those

in use today. In addition to the technology evolution issue, the technical points listed below are also important when interpreting older work [(Geoffrey., et al 1993), (Ruzin., et al 1999). This, along with the developmental nature of most of the devices in earlier studies, makes it very difficult to compare older results with more contemporary work. Changes in device design and structure occurred very rapidly. It is often possible to adjust the earlier data using the more modern interpretation of displacement damage with the Non-ionizing Energy Loss (NIEL) concept. Dark current in devices biased during irradiation indicates degradation is more severe than predicted by (NIEL). Note the relationship between the dose in energy deposited per unit volume and the fluence in particles per cm for protons is given by (Xiaoli ., et al 1997):

$$\text{Dose (rad (SI))} = \text{LET} \times \text{Fluence} \times 1.6 \times 10^{-8} \text{ rad (SI).g / MeV} \quad (1)$$

Where LET is the linear energy transfer coefficient in silicon and they are equal to 0.0578, 0.0239, and 0.0180 MeV cm/mg for 5.1, 16.2, and 23.4 MeV protons, respectively. APD technologies that are commonly found in commercially available devices since they are the most practical option for nuclear applications designers.

In the present study, we have presented the main irradiation effects, i.e., the multiplication gain, minority carrier life time, impact ionization, illumination and radiation damage coefficient. By comparing neutrons, protons and gamma radiation effects, we will apply the model on the two different Silicon avalanche photodiode structures. The results demonstrate that the model can accurately calculate the internal parameters of the APDs and produce data that can be directly compared with measurements. The fluence effects of proton irradiation on the photosensitivity and signal to noise ratio (SNR) are also investigated. The objective was to analyze the effect depletion region volume, Carrier concentration of the i-region of APDs on radiation hardness.

## APD DEVICE MODELING ANALYSIS

The dark current,  $I_{\text{Dark}}$ , for a device having depletion depth  $W$ , active area  $A$  and the effective carrier concentration,  $N_{\text{eff}}$  under high temperature irradiation  $T$  and the radiation fluence  $\Phi$  is given by (Srour., et al 2000):

$$I_{\text{Dark}} = \frac{qAW(T, \phi)N_{\text{eff}}(T, \phi)}{2\tau_r(T, \phi)} \quad (2)$$

Where  $\tau_r$  is the minority carrier lifetime after irradiation and it is given by (Osborne., et al 2000):

$$1/\tau_r = 1/\tau_0 + K_r\phi \quad (3)$$

Where  $\tau_0$  denotes the pre-irradiation minority carrier lifetime and  $K_r$  is the damage coefficient for  $\tau_r$ . For silicon detectors, changes in dark current per unit depletion region volume have been expressed as (Becker., et al 2003):

$$\frac{\Delta I_{\text{Dark}}}{V_V} = \frac{qN_{\text{eff}}}{2K_m} \phi \quad (4)$$

where  $V_V$  is the depletion region volume, and  $K_m$  is the damage coefficient for the material type in the depletion region. A simple model of the annealing can be constructed if we assume that the radiation-induced defects anneal according to a first-order mechanism (Srouf., et al 2000), at a given absolute high temperature irradiation  $T$ ,  $K_m$  can be related to an activation energy  $E$  by the Arrhenius formula:

$$K_m(T) = K_m(0) \exp(E / K_B T) \quad (5)$$

Where  $K_B$ , is Boltzmann's constant.

Also, defects generated during irradiation cause changes in the effective substrate doping concentration and consequently in the depletion voltage,  $V$ . The two microscopic mechanisms related to the  $N_{eff}$  variation are the donor removal (in n-type silicon) and the deep acceptor level generation, which are macroscopically modeled as a function of the radiation particle fluence,  $\phi$  by [6, 26]:

$$N_{eff}(\phi) = N_0 \exp(-c \phi) - \beta \phi \quad (6)$$

Where  $N_0$  is the donor concentration before irradiation,  $c$  is the donor removal coefficient, and  $\beta$  is the acceptor introduction rate. By increasing  $\phi$ ,  $N_{eff}$  decreases. The depletion width  $W$  can be expressed as the following (Andrea ., et al 2003), (McPherson, 2005):

$$W = \sqrt{\frac{2 \varepsilon (|V| + V_{bi})}{q N_{eff}}} \quad (7)$$

Where  $q$  is the electronic charge,  $\varepsilon$  is the absolute silicon dielectric constant and  $V_{bi} \approx 0.6$  Volt is the junction built-in potential. In order to analyze the response time of irradiated PIN photodiode, assume a modulated photon flux density as:

$$\phi = \phi_0 \exp(j \omega t) \text{ photons } / (s.cm^2) \quad (8)$$

To fall on photodiode, where  $\omega$  is the sinusoidal modulation frequency. The total current density through the depletion region generated by this photon flux can be shown to be (Sze, 2007):

$$J_{photo} = \left( \frac{j \omega \varepsilon (|V| + V_{bi})}{W} + q \phi_0 \frac{1 - e^{-j \omega t_{dr}}}{j \omega t_{dr}} \right) e^{j \omega t} \quad (9)$$

$$\left| \frac{J_{photo}}{a q \phi_0} \right| = \sqrt{\frac{\text{Sin}^2 \left( \frac{\omega t_{dr}}{2} \right) \left( 1 - \frac{\omega \varepsilon (|V| + V_{bi})}{W (\omega t_{dr})^2} \right) + \left( \frac{\omega \varepsilon (|V| + V_{bi})}{W} \right)^2}{\left( \frac{\omega t_{dr}}{2} \right)^2}} \quad (10)$$

Where  $a$  is the photodiode area,  $\varepsilon$  is the material permittivity and  $t_{dr}$  is the transit drift time of carriers through the depletion region. The time for diffusion of carriers from the undepleted region to the depleted region is given by:

$$t_{df} = \frac{\ell^2}{2D} \quad (11)$$

Where  $D$  and  $\ell$  are the diffusion constant and the undepleted thickness, which changes with the changing of the depletion layer width  $W$ , since  $\ell = W_0 - w$ ,  $W_0$  is the substrate thickness. The diffusion current arises from the

regions within a diffusion length of the minority carriers next to the junction:

$$D = \frac{L_p^2}{\tau_r} \quad (12)$$

The irradiation induce diffusion length change can be expressed as the following (Abd El-Naser ., 2009):

$$L_p = \left( \frac{1}{\alpha} \right) \frac{1 - e^{\alpha W} \left( 1 - \frac{I_{photo}}{a q \phi_0} \right)}{e^{\alpha W} \left( 1 - \frac{I_{photo}}{a q \phi_0} \right)} \quad (13)$$

Where  $\alpha$  is the absorption coefficient. The time constant  $t_{RC}$  of the photodiode with a load resistance  $R_L$  is given by:

$$t_{RC} = 2.2(R_S + R_L)C \quad (14)$$

Where  $C$  is the capacitance of photodiode,  $R_S$  is the series resistance of photodiode and it is given by:

$$R_S = \frac{\rho (W_0 - W)}{A} + \text{contact resistore} \quad (15)$$

Where  $\rho$  is resistivity of the substrate. Finally, for fully depleted photodiodes the rise time  $t_r$  and fall time are generally the same.

$$t_r = \sqrt{t_{dr}^2 + t_{df}^2 + t_{RC}^2} \quad (16)$$

Radiation induced changes in dark current are important to quantify, because dark current changes are an important component of such figures of merit as signal to noise ratio and noise equivalent power (NEP). The total spectral noise current of an unilluminated APD is given by:

$$I_n = \sqrt{2 q (I_{ds} + I_{db} M^2 F) B} \quad (17)$$

Where  $I_{ds}$  is the unmultiplied surface dark current,  $I_{db}$  is the gain-multiplied bulk dark current,  $M$  is the gain,  $F$  is the excess noise factor, and  $B$  is the noise bandwidth [1]. Total dark current ( $I_{Dark}$ ) is related to the parameters in (2) by:

$$I_{Dark} = I_{ds} + M I_{db} \quad (18)$$

$$M I_{db} \approx I_{Dark} \gamma - I_{Dark} \text{ protons} \approx I_{Dark} \text{ neutrons} \quad (19)$$

The gain of an APD can be easily measured by continuous light method: the dark current and the current under continuous illumination are recorded for each fluence. The gain is then calculated as the current amplification with respect to a reference bias, where no amplification is assumed:

$$M(\phi, V_r) = \frac{I_{Photo}(\phi, V_r) - I_{Dark}(\phi, V_r)}{I_{Photo}(0, V_r) - I_{Dark}(0, V_r)} \quad (20)$$

The gain as a function of bias voltage at different radiation fluence, it is usually described by the formula:

$$M(V) = \frac{1}{1 - \left( \frac{V}{V_b} \right)^n} \quad (21)$$

Where  $V_b$  is the breakdown voltage and  $n$  is a coefficient to be determined experimentally. Both  $V_b$  and  $n$  depend on the fluence.

$$\frac{1}{M} \frac{dM}{d\phi} = \frac{1}{M} \frac{\partial M}{\partial V_b} \frac{\partial V_b}{\partial \phi} + \frac{1}{M} \frac{\partial M}{\partial n} \frac{\partial n}{\partial \phi} = -M \left( \frac{V}{V_b} \right)^n \left[ \frac{n}{V_b} \frac{\partial V_b}{\partial \phi} + L_n \left( \frac{V}{V_b} \right) \frac{\partial n}{\partial \phi} \right] \quad (22)$$

**Table 1.** Proposed operating parameters for our suggested APD device.

Operating parameter	Symbol	Value
Radiation dose or fluence	$\Phi$	$3 \times 10^{14}$ p/cm <sup>2</sup> – $5 \times 10^{15}$ p/cm <sup>2</sup>
Thermal activation energy	E	0.4 eV–1.0 eV
The amplifier noise figure	$F_n$	2 dB
n coefficient	n	1–2
Boltzman's constant	$K_B$	$1.38 \times 10^{-23}$ J/K
Donor concentration before irradiation	$N_0$	$10^{17}$ cm <sup>-3</sup> – $10^{20}$ cm <sup>-3</sup>
Acceptor introduction rate	B	0.0205–0.0248 cm <sup>-1</sup>
Donor removal coefficient	C	0.0008–0.002 cm <sup>-1</sup>
Ambient temperature	T	280 K–340 K
Initial output power	$P_0$	0.1 mWatt–0.497 mWatt
The absorption coefficient	$\alpha$	$10^3$ cm <sup>-1</sup> – $10^4$ cm <sup>-1</sup>
Initial multiplication gain	M	20-150
Depletion region voltage	V	1 Volt–10 Volt
Bias voltage	$V_{bias}$	50–400 Volt
Effective ionization rate	K	0.015–0.035
Electron charge	q	$1.6 \times 10^{-19}$ J/eV
Initial carrier life time	$T_0$	2 nsec–10 nsec
Optical Signal wavelength	$\lambda$	0.65-1-1.3 $\mu$ m
Quantum efficiency	$\eta$	75%–90%

In the present analysis, we assume the optical signal power to be intensity modulated at a sinusoidal modulation frequency  $\omega$  such that yields:

$$P(t) = P_0(1 + m \exp(j\omega t)) \tag{23}$$

Where  $P_0$  is the average optical power signal density and m, the index of modulation. The frequency dependent multiplication gain has been computed from (Jamie ., et al 2006), (Jamie ., et al 2006):

$$M(\omega) = \frac{M}{\sqrt{1 + \alpha(Mt_{rc} + t_{td})^2}} \tag{24}$$

Where  $t_r$  is the intrinsic response time and M, the multiplication gain. For the APD structure, The excess noise factor F depends on the type of primary injection (electron, hole or mixed injection) and ionization rates. The excess noise factor is given by (Pilicer., et al 2005), (Jamie ., et al 2003):

$$F(M) = MK + \left[ 2 - \frac{1}{M} \right] (1 - K) \tag{25-a}$$

$$F(M) = M \left[ 1 - (1 - K) \left( \frac{M-1}{M} \right)^2 \right] \tag{25-b}$$

From experimental observations it has been found that, in general, the excess noise factor can be approximated by:

$$F(M) = M^x \tag{26}$$

Where K is the effective ionization rate ratio, the parameter x take on values of 0 for PIN photodiode and 0.3 for Si, 0.7 for InGaAs avalanche photodiode. Where the quantum efficiency,  $\eta$ , can be given by:

$$\eta = \frac{I_{photo} / q}{P_0(1 - r) / h\nu} \tag{27}$$

The multiplication mechanism in the avalanche region multiplies the background current, signal current and dark current. For the modulated signal with average power density  $P_0$ , the signal-to-noise power ratio of the APD can be obtained as the following expression:

$$S / N = \frac{0.5m^2 I_{photo}^2}{2qMFB(I_{photo} + I_{Dark}) + \frac{4K_B T B F_n}{R_L M^2}} \tag{28}$$

Where  $B \approx 0.35/t_r$  is bandwidth and  $t_r$  is rise time. The term  $(4K_B T B F_n / R_L)$  is the total noise associated with amplifier, it is referred to thermal noise of load resistor  $R_L$  by the amplifier noise figure  $F_n$ . The optimum value of multiplication gain for the maximum signal-to-noise ratio can be obtained by setting the first derivative of S/N ratio with gain to zero. This yields:

$$M^3 q (I_{photo} + I_{Dark}) R_L \left[ 1 + \left( \frac{M^2 - 1}{M} \right) \left( K + \frac{\partial K}{\partial \phi} \frac{\partial \phi}{\partial M} - 1 \right) \right] - 4kT = 0 \tag{29}$$

Doing so for a sinusoidally modulated signal, with  $m=1$  and F approximated by  $M^x$ , yields:

$$M^{x+2} = \frac{2qI_{Dark} + 4K_B T / R_L}{xq(I_{photo} + I_{Dark})} \tag{30}$$

The noise equivalent power (NEP) of the APD which corresponds to the rms value of the incident optical power required to produce unity signal-to-noise ratio in a 1 Hz bandwidth has been computed from

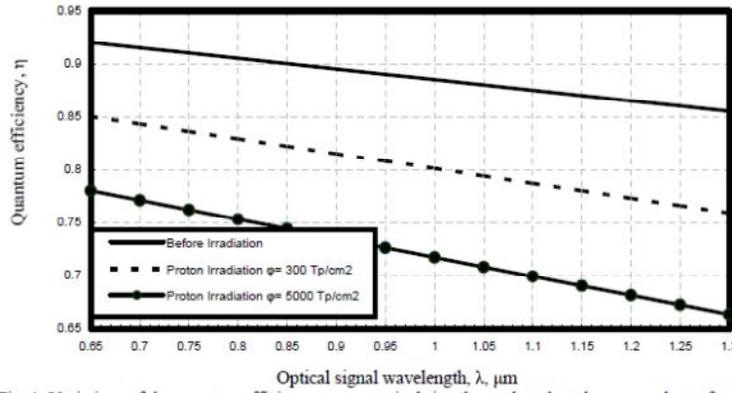


Fig. 1. Variations of the quantum efficiency versus optical signal wavelength at the assumed set of parameters.

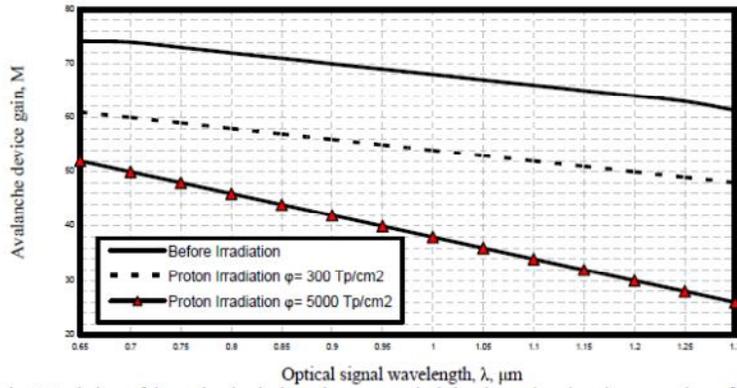


Fig. 2. Variations of the avalanche device gain versus optical signal wavelength at the assumed set of parameter

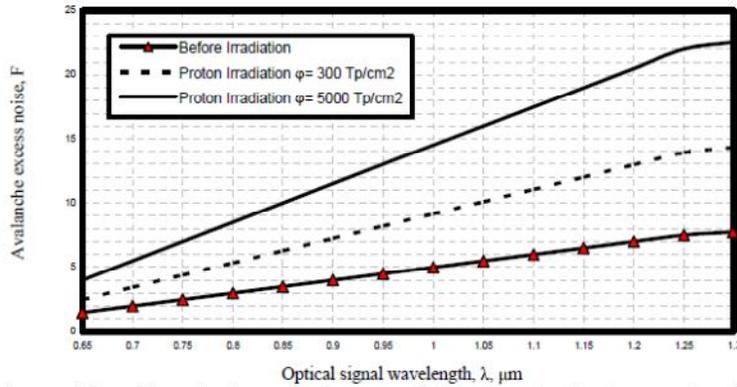


Fig. 3. Variations of the avalanche excess noise versus optical signal wavelength at the assumed set of parameter

$$NEP = \frac{\sqrt{2}h\nu}{\eta} \left[ \frac{qR_L M^3 (I_{Photo} + I_{Dark}) \left( 1 - (1-K) \left( \frac{M-1}{M} \right)^2 \right) + 2KT}{q^2 R_L M^4 \left( 1 - (1-K) \left( \frac{M-1}{M} \right)^2 \right)} \right]^{0.5} \quad (31)$$

In digital optical communication, the bit error rate (BER) for the APD can be written as:

$$BER = \frac{1}{2} \operatorname{erfc} \left[ \frac{\left( \frac{q\eta\mu_{opt}AM}{h\nu} \right)^2 R_L}{4\sqrt{2} \left[ 2qR_L M^3 (I_{Photo} + I_{Dark}) \left( 1 - (1-K) \left( \frac{M-1}{M} \right)^2 \right) + 4KT \right]} \right] \quad (32)$$

## SIMULATION RESULTS AND DISCUSSIONS

We have investigated deeply avalanche photodiodes as a media of optical detectors under thermal irradiated fields under the assumed set of the operating parameters listed in Table 1.

Based on the model equations analysis, assumed set of the operating parameters, and the set of the Figs. (1-12), the following facts are assured as in the following results:

- i) As shown in Figs. (1, 2) have assured that as both optical signal wavelength and proton irradiation

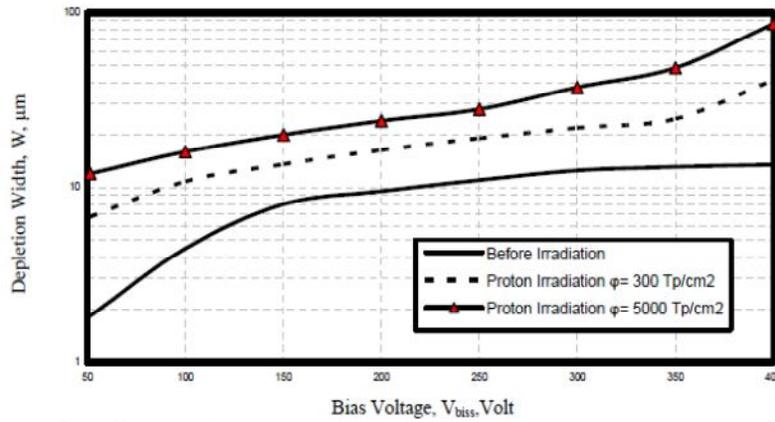


Fig. 4. Variations of the depletion width against bias voltage at the assumed set of parameters.

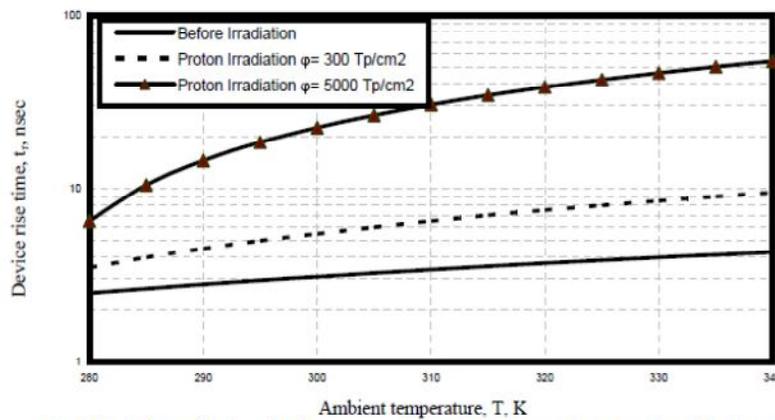


Fig. 5. Variations of device rise time versus ambient temperature at the assumed set of parameters.

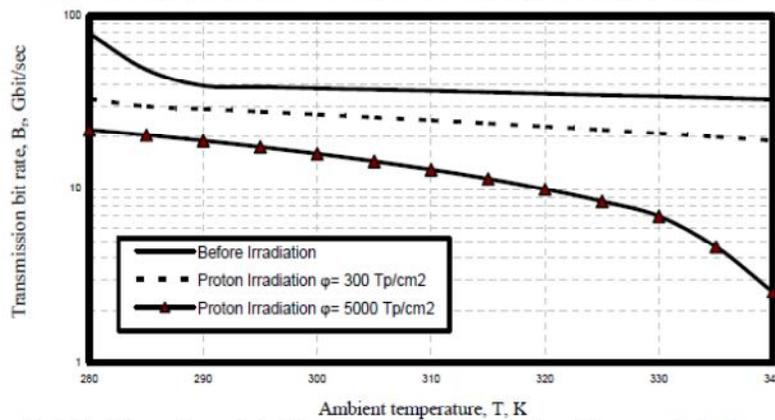


Fig. 6. Variations of transmission bit rate against ambient temperature at the assumed set of parameters.

fluences increase, this leads to decrease in both device gain and quantum efficiency.

ii) Fig. 3 has demonstrated that as both optical signal wavelength and proton irradiation fluences increase, this results in increasing of excess noise to the device. Fig. 4 has proved that as both bias voltage and proton irradiation fluences increase, this leads to increase in depletion width in the junction devices area.

iii) As shown in Fig. 5 has indicated that as both ambient temperature and proton irradiation fluences increase, this results in increasing of rise time of the device pulse.

iv) As shown in Figs. (6, 7) have indicated that as both ambient temperature and proton irradiation fluences increase, this results in decreasing of both signal to noise ratio and transmission bit rates.

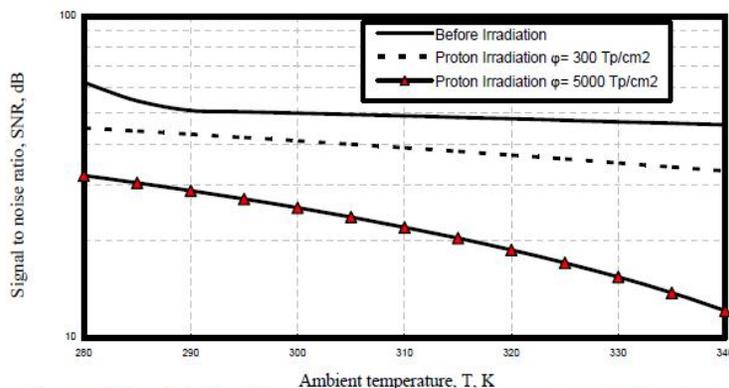


Fig. 7. Variations of signal to noise ratio against ambient temperature at the assumed set of parameters.

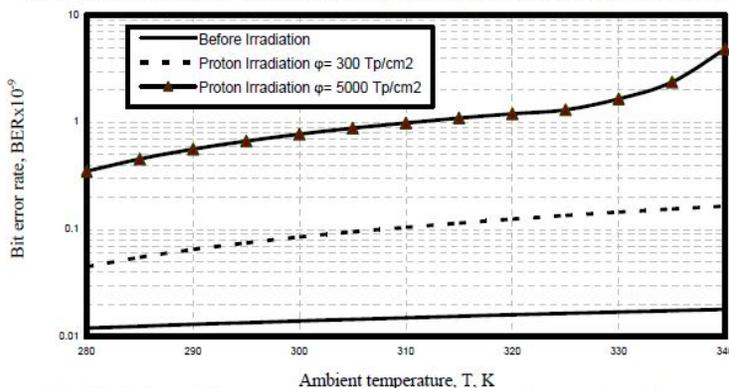


Fig. 8. Variations of bit error rate against ambient temperature at the assumed set of parameters.

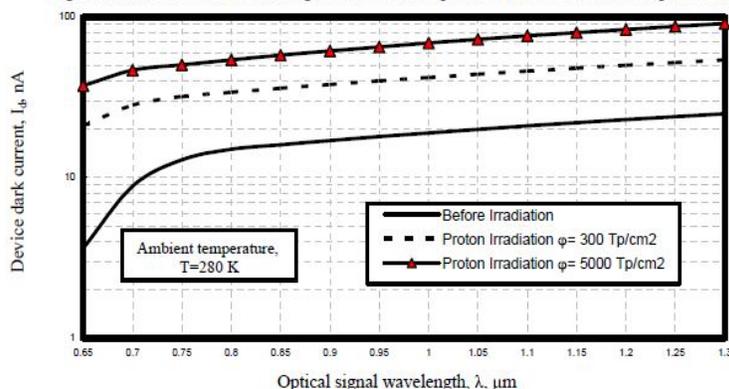


Fig. 9. Variations of the device dark current against optical signal wavelength at the assumed set of parameters.

- v) As shown in Fig. 8 has proved that as both ambient temperature and proton irradiation fluences increase, this results in increasing of bit error rate at the receiving side.
- vi) Figs. (9-11) have assured that as ambient temperature, optical signal wavelength, and proton irradiation fluences increase, this leads to increase of device dark current.
- vii) As shown in Fig. 12 has demonstrated that as both ambient temperature, and proton irradiation fluences increase, this leads to increase of noise equivalent power of the avalanche photodiode device.

## CONCLUSIONS

In a summary, We have been developed the transmission performance characteristics of avalanche photodiode devices under higher temperature effects and within proton irradiation fluences environments. It is theoretically found that the increased operating optical signal wavelength and proton irradiation fluences, this results in the increased excess noise and the decreased device quantum efficiency and device gain. As well as the increased ambient temperature and proton irradiation fluences, this leads to the decreased of both signal to

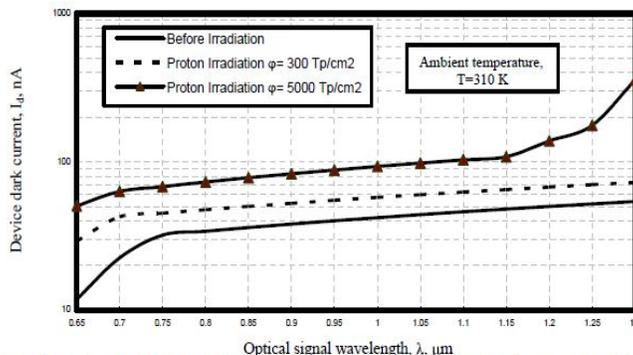


Fig. 10. Variations of the device dark current against optical signal wavelength at the assumed set of parameters.

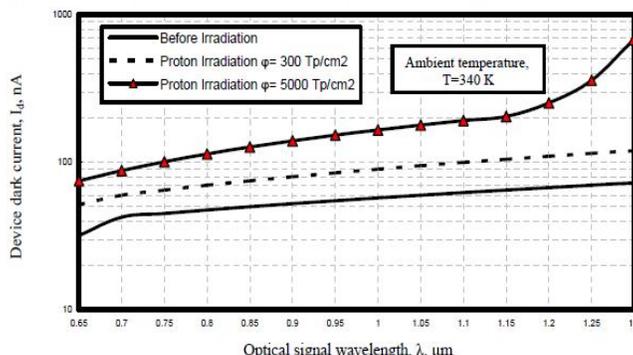
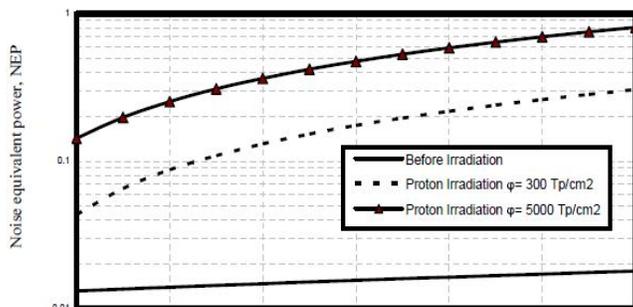


Fig. 11. Variations of the device dark current against optical signal wavelength at the assumed set of parameters.



noise ratio and transmission bit rates, and the increased of both bit error rate and pulse rise time. It is evident that the increased ambient temperature, operating optical signal wavelength, and proton irradiation fluences, this results in the increased of both dark current and noise equivalent power.

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